Analysis of the fractional descriptor linear systems by the decomposition into the dynamical and static parts

Tadeusz KACZOREK D and Lukasz SAJEWSKI

A new method of the decomposition of the fractional descriptor linear continuoustime and discrete-time systems into dynamical and static parts is proposed. Conditions for the decomposition of the fractional descriptor linear systems are established and procedures for compositions of the matrices of dynamical and static parts are given. The procedures are illustrated by numerical examples.

Key words: decomposition, descriptor, fractional, linear, procedure, system, dynamical, static, part

1. Introduction

Mathematical fundamentals of the fractional calculus are given in the monographs [6, 8, 11, 12]. The descriptor linear systems have been analyzed in many books [1, 2, 6–8] and the fractional standard and descriptor linear systems in [9, 11, 12, 14–16]. Positive linear systems consisting of *n* subsystems with different fractional orders have been introduced and investigated in [3, 5–8, 10]. The fractional descriptor discrete-time linear systems have been analyzed by the use of the shuffle algorithm in [9]. The stability of the delayed fractional discrete-time linear system has been investigated in [13, 14]. The decentralized stabilization of descriptor fractional positive discrete-time linear systems with delays have been addressed in [15] and the stabilization of positive descriptor

Copyright © 2024. The Author(s). This is an open-access article distributed under the terms of the Creative Commons Attribution-NonCommercial-NoDerivatives License (CC BY-NC-ND 4.0 https://creativecommons.org/licenses/ by-nc-nd/4.0/), which permits use, distribution, and reproduction in any medium, provided that the article is properly cited, the use is non-commercial, and no modifications or adaptations are made

T. Kaczorek (e-mail: t.kaczorek@pb.edu.pl) and L. Sajewski (corresponding author, e-mail: l.sajewski @pb.edu.pl) are with Faculty of Electrical Engineering, Bialystok University of Technology, Wiejska 45D, 15-351 Białystok, Poland.

The studies have been carried out in the framework of work No. WZ/WE-IA/5/2023 financed from the funds for science by the Polish Ministry of Science and Higher Education.

Received 6.6.2024. Revised 17.10.2024.

fractional continuous-time linear system with two different fractional orders by decentralized controller in [16].

In this paper a method of the decomposition of the fractional linear systems into the dynamical and static parts is proposed.

The paper is organized as follows. In Section 2 the decomposition method for the fractional descriptor linear continuous-time systems is proposed. The extension of this method to the fractional linear discrete-time systems is presented in Section 3. Concluding remarks are given in Section 4.

The following notation will be used: \Re – the set of real numbers, $\Re^{n \times m}$ – the set of $n \times m$ real matrices, I_n – the $n \times n$ identity matrix.

2. Decomposition of the fractional descriptor linear continuous-time systems

Consider the fractional linear system

$$E\frac{\mathrm{d}^{\alpha}x}{\mathrm{d}t^{\alpha}} = Ax + Bu, \quad 0 < \alpha < 1, \tag{1a}$$

$$y = Cx, \tag{1b}$$

where $x = x(t) \in \mathbb{R}^n$, $u = u(t) \in \mathbb{R}^m$, $y = y(t) \in \mathbb{R}^p$ are the state, input and output vectors, $E, A \in \mathbb{R}^{n \times n}, B \in \mathbb{R}^{n \times m}, C \in \mathbb{R}^{p \times n}$ and

$$\frac{\mathrm{d}^{\alpha}x(t)}{\mathrm{d}t^{\alpha}} = \frac{1}{\Gamma(1-\alpha)} \int_{0}^{t} \frac{\dot{x}(\tau)}{(t-\tau)^{\alpha}} \mathrm{d}\tau, \quad \dot{x}(\tau) = \frac{\mathrm{d}x(\tau)}{\mathrm{d}\tau} \tag{1c}$$

is the Caputo fractional derivative and

$$\Gamma(z) = \int_{0}^{\infty} t^{z-1} e^{-t} dt, \quad \text{Re}(z) > 0$$
 (1d)

is the gamma function [6-8, 12].

It is assumed that

det
$$E = 0$$
 and det $[Es^{\alpha} - A] \neq 0.$ (2)

In this case equation (1a) has unique solution.

The following elementary operations on real matrices will be used [7]:

1. Multiplication of any *i*-th row (column) by the number *a*. This operation will be denoted by $L[i \times a]$ for row operation and by $R[i \times a]$ for column operation.

- 2. Addition to any *i*-th row (column) of the *j*-th row (column) multiplied by any number *b*. This operation will be denoted by $L[i + j \times b]$ for row operation and by $R[i + j \times b]$ for column operation.
- 3. L[i, j] for interchange of rows and R[i, j] for interchange of columns.

The elementary operations do not change the rank of the matrices [7].

Using the row elementary operations to the equation (1a) we obtain

$$\begin{bmatrix} E_1 \\ 0 \end{bmatrix} \frac{\mathrm{d}^{\alpha} x}{\mathrm{d} t^{\alpha}} = \begin{bmatrix} A_1 \\ A_2 \end{bmatrix} x + \begin{bmatrix} B_1 \\ B_2 \end{bmatrix} u, \tag{3a}$$

where E_1 , $A_1 \in \mathfrak{R}^{r \times n}$, $B_1 \in \mathfrak{R}^{r \times m}$, $A_2 \in \mathfrak{R}^{(n-r) \times n}$, $B_2 \in \mathfrak{R}^{(n-r) \times m}$ and

$$\operatorname{rank} E_1 = \operatorname{rank} E = r < n. \tag{3b}$$

Applying the column elementary operations to the matrices $\begin{bmatrix} E_1 \\ 0 \end{bmatrix}$, $\begin{bmatrix} A_1 \\ A_2 \end{bmatrix}$ we obtain

$$\begin{bmatrix} E_{11} & E_{12} \\ 0 & 0 \end{bmatrix} \frac{\mathrm{d}^{\alpha}}{\mathrm{d}t^{\alpha}} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} B_1 \\ B_2 \end{bmatrix} u, \tag{4}$$

where det $E_{11} \neq 0$ and det $A_{22} \neq 0$ since (2).

From

$$A_{21}x_1 + A_{22}x_2 + B_2u = 0 \tag{5}$$

we have

$$x_2 = -A_{22}^{-1} \left(A_{21} x_1 + B_2 u \right) \tag{6a}$$

and

$$\frac{\mathrm{d}^{\alpha} x_2}{\mathrm{d} t^{\alpha}} = -A_{22}^{-1} \left(A_{21} \frac{\mathrm{d}^{\alpha} x_1}{\mathrm{d} t^{\alpha}} + B \frac{\mathrm{d}^{\alpha} u}{\mathrm{d} t^{\alpha}} \right). \tag{6b}$$

Substituting (6b) into (4) we obtain

$$\begin{bmatrix} \bar{E}_{11} & 0\\ \bar{A}_{21} & I_{n-r} \end{bmatrix} \frac{\mathrm{d}^{\alpha}}{\mathrm{d}t^{\alpha}} \begin{bmatrix} x_1\\ x_2 \end{bmatrix} = \begin{bmatrix} \bar{A}_{11} & 0\\ 0 & 0 \end{bmatrix} \begin{bmatrix} x_1\\ x_2 \end{bmatrix} + \begin{bmatrix} \bar{B}_{10}\\ 0 \end{bmatrix} u + \begin{bmatrix} \bar{B}_{11}\\ \bar{B}_{21} \end{bmatrix} \frac{\mathrm{d}^{\alpha}u}{\mathrm{d}t^{\alpha}}, \quad (7a)$$

where

$$\bar{E}_{11} = E_{11} - E_{12}A_{22}^{-1}A_{21},
\bar{A}_{11} = A_{11} - A_{12}A_{22}^{-1}A_{21}, \quad \bar{A}_{21} = A_{22}^{-1}A_{21},
\bar{B}_{10} = B_1 - A_{12}A_{22}^{-1}B_2, \quad \bar{B}_{11} = E_{12}A_{22}^{-1}B_2, \quad \bar{B}_{21} = -A_{22}^{-1}B_2.$$
(7b)

Lemma 1. The matrix

$$\bar{E}_{11} = E_{11} - E_{12} A_{22}^{-1} A_{21} , \qquad (8)$$

is nonsingular if the assumptions (2) are satisfied.

Proof. From assumption (2) it follows that

$$\det \begin{bmatrix} E_{11} & E_{12} \\ A_{21} & A_{22} \end{bmatrix} \neq 0.$$
(9)

From the equality

$$\begin{bmatrix} I_r & -E_{12}A_{22}^{-1} \\ 0 & I_{n-r} \end{bmatrix} \begin{bmatrix} E_{11} & E_{12} \\ A_{21} & A_{22} \end{bmatrix} = \begin{bmatrix} E_{11} - E_{12}A_{22}^{-1}A_{21} & 0 \\ A_{21} & A_{22} \end{bmatrix}$$
(10)

we have det $\overline{E}_{11} \neq 0$ since (9) holds and det $A_{22} \neq 0$.

Considering that det $\bar{E}_{11} \neq 0$ from (7a) we obtain

$$\frac{\mathrm{d}^{\alpha}}{\mathrm{d}t^{\alpha}} \begin{bmatrix} x_1\\ x_2 \end{bmatrix} = \begin{bmatrix} \hat{A}_{11} & 0\\ \hat{A}_{21} & 0 \end{bmatrix} \begin{bmatrix} x_1\\ x_2 \end{bmatrix} + \begin{bmatrix} \hat{B}_{10}\\ \hat{B}_{20} \end{bmatrix} u + \begin{bmatrix} \hat{B}_{11}\\ \hat{B}_{21} \end{bmatrix} \frac{\mathrm{d}^{\alpha}u}{\mathrm{d}t^{\alpha}}$$
(11a)

where

$$\hat{A}_{11} = \bar{E}_{11}^{-1} \bar{A}_{11}, \quad \hat{A}_{21} = -\bar{A}_{21} \bar{E}_{11}^{-1} \bar{A}_{11},$$

$$\hat{B}_{10} = \bar{E}_{11}^{-1} \bar{B}_{10}, \quad \hat{B}_{20} = -\bar{A}_{21} \bar{E}_{11}^{-1} \bar{B}_{10},$$

$$\hat{B}_{11} = \bar{E}_{11}^{-1} \bar{B}_{11}, \quad \hat{B}_{21} = \bar{B}_{21} - \bar{A}_{21} \bar{E}_{11}^{-1}.$$
(11b)

The equation (11a) describes the dynamical part of the descriptor system and the equation (5) the static part.

Therefore, the following theorem has been proved.

Theorem 1. *The fractional descriptor system* (1) *can be decomposed into its dynamical part* (11) *and its static part* (5).

To compute the matrices of the dynamical and static parts of the fractional descriptor system the following procedure can be used.

Procedure 1.

- **Step 1.** Check the conditions (2).
- **Step 2.** Using elementary column operations to (1a) find (3) and the matrices E_1, A_1, A_2, B_1, B_2 .
- **Step 3.** Using (7b) compute the matrices \bar{E}_{11} , \bar{A}_{11} , \bar{A}_{21} , \bar{B}_{10} , \bar{B}_{11} , \bar{B}_{21} .
- **Step 4.** Using (11b) compute the matrices \hat{A}_{11} , \hat{A}_{21} , \hat{B}_{10} , \hat{B}_{20} , \hat{B}_{11} , \hat{B}_{21} of the dynamical part. The matrices A_{21} , A_{22} , B_2 of the static part are given (5).

The procedure will be illustrated by the following numerical example.

Example 1. Consider the fractional descriptor linear system (1) with the matrices

$$E = \begin{bmatrix} 1 & 0 & 0 & 0 \\ -2 & 1 & 0 & 1 \\ -2 & -1 & 0 & -1 \\ 2 & 0 & 0 & 0 \end{bmatrix}, \quad A = \begin{bmatrix} 0 & 1 & 2 & 1 \\ 1 & 2 & 4 & 1 \\ -1 & 0 & -4 & 0 \\ 0 & 2 & 5 & 2 \end{bmatrix}, \quad B = \begin{bmatrix} 1 & 0 \\ 2 & 0 \\ -1 & 0 \\ 2 & 1 \end{bmatrix}.$$
(12)

Using Procedure 1 we obtain

Step 1. In this case the conditions (2) are satisfied since

$$\det E = \begin{vmatrix} 1 & 0 & 0 & 0 \\ -2 & 1 & 0 & 1 \\ -2 & -1 & 0 & -1 \\ 2 & 0 & 0 & 0 \end{vmatrix} = 0$$
(13a)

and

$$\det[Es^{\alpha} - A] = \begin{vmatrix} s^{\alpha} & -1 & -2 & -1 \\ -2s^{\alpha} - 1 & s^{\alpha} - 2 & -4 & s^{\alpha} - 1 \\ -2s^{\alpha} + 1 & -s^{\alpha} & 4 & -s^{\alpha} \\ 2s^{\alpha} & -2 & -5 & -2 \end{vmatrix} = s^{2\alpha} + 2s^{\alpha} - 1.$$
(13b)

Step 2. Applying the elementary row operation: $L[2+1\times(-1)], L[3+2\times(-1)], L[4+1\times2]$ to the matrices

$$E, A, B = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 1 & 2 & 1 & 1 & 0 \\ -2 & 1 & 0 & 1 & 1 & 2 & 4 & 1 & 2 & 0 \\ -2 & -1 & 0 & -1 & -1 & 0 & -4 & 0 & -1 & 0 \\ 2 & 0 & 0 & 0 & 0 & 2 & 5 & 2 & 2 & 1 \end{bmatrix}$$
(14)

we obtain

$$\begin{bmatrix} E_{11} & E_{12} & A_{11} & A_{12} & B_1 \\ 0 & 0 & A_{21} & A_{22} & B_2 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 & 0 & 0 & 1 & 2 & 1 & 1 & 0 \\ 0 & 1 & 0 & 1 & 1 & 0 & 0 & -1 & 0 & 0 \\ 0 & 0 & 0 & 0 & 0 & 2 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 0 & 1 \end{bmatrix}.$$
 (15)

Step 3. Using (7b) and (15) we obtain

$$\bar{E}_{11} = E_{11} - E_{12}A_{22}^{-1}A_{21} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} - \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} 0 & 2 \\ 0 & 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0 & -1 \end{bmatrix},
\bar{A}_{11} = A_{11} - A_{12}A_{22}^{-1}A_{21} = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} - \begin{bmatrix} 2 & 1 \\ 0 & -1 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} 0 & 2 \\ 0 & 0 \end{bmatrix} = \begin{bmatrix} 0 & -1 \\ 1 & 2 \end{bmatrix},
\bar{A}_{21} = A_{22}^{-1}A_{21} = \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} 0 & 2 \\ 0 & 0 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 0 & 2 \end{bmatrix},
\bar{B}_{10} = B_1 - A_{12}A_{22}^{-1}B_2 = \begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix} - \begin{bmatrix} 2 & 1 \\ 0 & -1 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} = \begin{bmatrix} 0 & -2 \\ 1 & 0 \end{bmatrix},
\bar{B}_{11} = E_{12}A_{22}^{-1}B_2 = \begin{bmatrix} 0 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix},
\bar{B}_{21} = -A_{22}^{-1}B_2 = -\begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix}^{-1} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} = \begin{bmatrix} 0 & -1 \\ -1 & 0 \end{bmatrix}.$$
(16)

Step 4. Using (11b) and (16) we obtain

$$\hat{A}_{11} = \bar{E}_{11}^{-1} \bar{A}_{11} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}^{-1} \begin{bmatrix} 0 & -1 \\ -1 & 2 \end{bmatrix} = \begin{bmatrix} 0 & -1 \\ 1 & -2 \end{bmatrix},$$

$$\hat{A}_{21} = -\bar{A}_{21} \bar{E}_{11}^{-1} \bar{A}_{11} = -\begin{bmatrix} 0 & 0 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}^{-1} \begin{bmatrix} 0 & -1 \\ -1 & 2 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ -2 & -2 \end{bmatrix},$$

$$\hat{B}_{10} = \bar{E}_{11}^{-1} \bar{B}_{10} = \begin{bmatrix} 1 & 0 \\ -1 & 1 \end{bmatrix}^{-1} \begin{bmatrix} 0 & -2 \\ 1 & 0 \end{bmatrix} = \begin{bmatrix} 0 & -2 \\ 1 & -2 \end{bmatrix},$$

$$\hat{B}_{20} = -\bar{A}_{21} \bar{E}_{11}^{-1} \bar{B}_{10} = -\begin{bmatrix} 0 & 0 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ -1 & 1 \end{bmatrix}^{-1} \begin{bmatrix} 0 & -2 \\ 1 & 0 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ -2 & 4 \end{bmatrix},$$

$$\hat{B}_{11} = \bar{E}_{11}^{-1} \bar{B}_{11} = \begin{bmatrix} 1 & 0 \\ -1 & 1 \end{bmatrix}^{-1} \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix} = \begin{bmatrix} 0 & 0 \\ 1 & 0 \end{bmatrix},$$

$$\hat{B}_{21} = \bar{B}_{21} - \bar{A}_{21} \bar{E}_{11}^{-1} = \begin{bmatrix} 0 & -1 \\ -4 & 0 \end{bmatrix} - \begin{bmatrix} 0 & 0 \\ 0 & 2 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ -1 & 1 \end{bmatrix}^{-1} = \begin{bmatrix} 0 & -1 \\ -3 & -2 \end{bmatrix}.$$

Therefore, the dynamical part of the fractional descriptor system with (12) is described by the differential equation

$$\frac{\mathrm{d}^{\alpha}}{\mathrm{d}t^{\alpha}} \begin{bmatrix} x_1\\ x_2 \end{bmatrix} = \begin{bmatrix} \hat{A}_{11} & 0\\ \hat{A}_{21} & 0 \end{bmatrix} \begin{bmatrix} x_1\\ x_2 \end{bmatrix} + \begin{bmatrix} \hat{B}_{10}\\ \hat{B}_{20} \end{bmatrix} u + \begin{bmatrix} \hat{B}_{11}\\ \hat{B}_{21} \end{bmatrix} \frac{\mathrm{d}^{\alpha}u}{\mathrm{d}t^{\alpha}}$$
$$= \begin{bmatrix} 0 & -1 & 0 & 0\\ -1 & 2 & 0 & 0\\ 0 & 0 & 0 & 0\\ -2 & -2 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1\\ x_2 \end{bmatrix} + \begin{bmatrix} 0 & -2\\ 1 & -2\\ 0 & 0\\ -2 & 4 \end{bmatrix} u + \begin{bmatrix} 0 & 0\\ 1 & 0\\ 0 & -1\\ -3 & -2 \end{bmatrix} \frac{\mathrm{d}^{\alpha}u}{\mathrm{d}t^{\alpha}} \qquad (18)$$

and the static part by the algebraic equation

$$A_{21}x_1 + A_{22}x_2 + B_2u = \begin{bmatrix} 0 & 2 \\ 0 & 0 \end{bmatrix} x_1 + \begin{bmatrix} 0 & 1 \\ 1 & 0 \end{bmatrix} x_2 + \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} u = \begin{bmatrix} 0 \\ 0 \end{bmatrix}.$$
 (19)

3. Decomposition of the fractional descriptor linear discrete-time systems

In this section the considerations concerning the decomposition of the fractional descriptor linear system into dynamical and static parts will be extended to the fractional descriptor discrete-time systems.

Consider the fractional descriptor discrete-time linear system [6, 8]

$$E\Delta^{\alpha} x_{i+1} = A x_i + B u_i, \quad i \in Z_+ = 0, 1, 2, \dots, \quad 0 < \alpha < 1,$$
(20)

where $x_i \in \mathfrak{R}^n$, $u_i \in \mathfrak{R}^m$, $y_i \in \mathfrak{R}^p$ are the state, input and output vectors and $E, A \in \mathfrak{R}^{n \times n}, B \in \mathfrak{R}^{n \times m}, C \in \mathfrak{R}^{p \times n}$ and

$$\Delta^{\alpha} x_{i} = \sum_{j=0}^{i} (-1)^{j} {\alpha \choose j} x_{i-j},$$

$${\alpha \choose j} = \begin{cases} 1 & \text{for } j = 0, \\ \frac{\alpha(\alpha-1)\dots(\alpha-j+1)}{j!} & \text{for } j = 1, 2, \dots \end{cases}$$
(21)

is the fractional α -order difference of x_i .

Substitution of (21) into (20) yields

$$x_{i+1} = A_{\alpha} x_i - \sum_{j=2}^{i+1} c_j x_{i-j+1}, \quad i \in \mathbb{Z}_+,$$
 (22a)

where

$$A_{\alpha} = A + I_n \alpha. \tag{22b}$$

It is assumed that the matrices E and A of the system (20) satisfy the conditions

$$\det E = 0 \quad \text{and} \quad \det[Ez - A] \neq 0. \tag{23}$$

Using the elementary row operations to the equation (20) we obtain

$$\begin{bmatrix} E_1 \\ 0 \end{bmatrix} \Delta^{\alpha} x_i = \begin{bmatrix} A_1 \\ A_2 \end{bmatrix} x_i + \begin{bmatrix} B_1 \\ B_2 \end{bmatrix} u_i, \quad i = 0, 1, \dots$$
(24a)

where E_1 , $A_1 \in \mathfrak{R}^{r \times n}$, $B_1 \in \mathfrak{R}^{r \times m}$, $A_2 \in \mathfrak{R}^{(n-r) \times n}$, $B_2 \in \mathfrak{R}^{(n-r) \times m}$ and

$$\operatorname{rank} E_1 = \operatorname{rank} E = r < n. \tag{24b}$$

Applying the elementary column operations to the matrices $\begin{bmatrix} E_1 \\ 0 \end{bmatrix}$, $\begin{bmatrix} A_1 \\ A_2 \end{bmatrix}$ we obtain

$$\begin{bmatrix} E_{11} & E_{12} \\ 0 & 0 \end{bmatrix} \Delta^{\alpha} \begin{bmatrix} x_{1i} \\ x_{2i} \end{bmatrix} = \begin{bmatrix} A_{11} & A_{12} \\ A_{21} & A_{22} \end{bmatrix} \begin{bmatrix} x_{1i} \\ x_{2i} \end{bmatrix} + \begin{bmatrix} B_1 \\ B_2 \end{bmatrix} u_i, \quad i = 0, 1, \dots$$
(25)

where det $E_{11} \neq 0$ and det $A_{22} \neq 0$ by the assumptions (23).

From (25) we have

$$A_{21}x_{1i} + A_{22}x_{2i} + B_2u_i = 0, \quad i = 0, 1, \dots,$$
(26)

$$x_{2i} = -A_{22}^{-1}(A_{21}x_{1i} + B_2u_i), \quad i = 0, 1, \dots$$
(27)

and

$$\Delta^{\alpha} x_{2i} = -A_{22}^{-1} (A_{21} \Delta^{\alpha} x_{1i} + B \Delta^{\alpha} u_i), \quad i = 0, 1, \dots$$
 (28)

Substituting (28) into (25) we obtain

$$\begin{bmatrix} \bar{E}_{11} & 0\\ \bar{A}_{21} & I_{n-r} \end{bmatrix} \Delta^{\alpha} \begin{bmatrix} x_{1i}\\ x_{2i} \end{bmatrix} = \begin{bmatrix} \bar{A}_{11} & 0\\ 0 & 0 \end{bmatrix} \begin{bmatrix} x_{1i}\\ x_{2i} \end{bmatrix} + \begin{bmatrix} \bar{B}_{10}\\ 0 \end{bmatrix} u_i + \begin{bmatrix} \bar{B}_{11}\\ \bar{B}_{21} \end{bmatrix} \Delta^{\alpha} u_i , \quad i = 0, 1, \dots,$$
(29a)

where

$$\bar{E}_{11} = E_{11} - E_{12}A_{22}^{-1}A_{21},
\bar{A}_{11} = A_{11} - A_{12}A_{22}^{-1}A_{21},
\bar{B}_{10} = B_1 - A_{12}A_{22}^{-1}B_2,
\bar{B}_{11} = E_{12}A_{22}^{-1}B_2,
\bar{B}_{21} = -A_{22}^{-1}B_2.$$
(29b)

Lemma 2. The matrix \overline{E}_{11} defined by (29b) is nonsingular if the assumption (23) is satisfied.

Proof. Proof is similar to the proof of Lemma 1.

Taking into account the det E = 0 from (29a) we obtain

$$\Delta^{\alpha} \begin{bmatrix} x_{1i} \\ x_{2i} \end{bmatrix} = \begin{bmatrix} \hat{A}_{11} & 0 \\ \hat{A}_{21} & 0 \end{bmatrix} \begin{bmatrix} x_{1i} \\ x_{2i} \end{bmatrix} + \begin{bmatrix} \hat{B}_{10} \\ \hat{B}_{20} \end{bmatrix} u_i + \begin{bmatrix} \hat{B}_{11} \\ \hat{B}_{21} \end{bmatrix} \Delta^{\alpha} u_i , \quad i = 0, 1, \dots,$$
(30a)

where

$$\hat{A}_{11} = \bar{E}_{11}^{-1} \bar{A}_{11}, \quad \hat{A}_{21} = -\bar{A}_{21} \bar{E}_{11}^{-1} \bar{A}_{11}, \\ \hat{B}_{10} = \bar{E}_{11}^{-1} \bar{B}_{10}, \quad \hat{B}_{20} = -\bar{A}_{21} \bar{E}_{11}^{-1} \bar{B}_{10}, \\ \hat{B}_{11} = \bar{E}_{11}^{-1} \bar{B}_{11}, \quad \hat{B}_{21} = \bar{B}_{21} - \bar{A}_{21} \bar{E}_{11}^{-1}.$$
(30b)

Equations (30a) describes the dynamical part of the descriptor system and the equation (26) its static part of the system.

Therefore, the following theorem has been proved.

Theorem 2. The fractional descriptor linear discrete-time system (20) can be decomposed into its dynamical part (30a) and its static part (26).

To compute the matrices of the dynamical and static parts of the fractional descriptor discrete-time system Procedure 2.1 (with some evident modifications) can be used.

Example 2. Consider the fractional descriptor linear system (20) with the matrices

$$E = \begin{bmatrix} 1 & 2 & 2 \\ 0 & 2 & 1 \\ 1 & 0 & 1 \end{bmatrix}, \quad A = \begin{bmatrix} 1 & 1 & 1 \\ 1 & 0 & 1 \\ 0 & 2 & 1 \end{bmatrix}, \quad B = \begin{bmatrix} 1 & 1 \\ 1 & 0 \\ 1 & 2 \end{bmatrix}.$$
 (31)

Using Procedure 2.1 we obtain

Step 1. The conditions (23) are satisfied since

$$\det E = \begin{vmatrix} 1 & 2 & 2 \\ 0 & 2 & 1 \\ 1 & 0 & 1 \end{vmatrix} = 0$$
(32a)

and

$$\det[Ez - A] = \begin{vmatrix} z - 1 & 2z - 1 & 2z - 1 \\ -1 & 2z & z - 1 \\ z & -2 & z - 1 \end{vmatrix} = z^2 - 1$$
(32b)

Step 2. Applying the elementary row operations: $L[1+2\times(-1)], L[3+1\times(-1)]$ to the matrices

$$E, A, B = \begin{bmatrix} 1 & 2 & 2 & 1 & 1 & 1 & 1 & 1 \\ 0 & 2 & 1 & 1 & 0 & 1 & 1 & 0 \\ 1 & 0 & 1 & 0 & 2 & 1 & 1 & 2 \end{bmatrix}$$
(33)

we obtain

$$\begin{bmatrix} E_{11} & E_{12} & A_{11} & A_{12} & B_1 \\ 0 & 0 & A_{21} & A_{22} & B_2 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 1 & 0 & 1 & 0 & 0 & 1 \\ 0 & 2 & 1 & 1 & 0 & 1 & 1 & 0 \\ 0 & 0 & 0 & 0 & 1 & 1 & 0 & 1 \end{bmatrix}.$$
 (34)

Step 3. Using (29b) and (34) we obtain

$$\bar{E}_{11} = E_{11} - E_{12}A_{22}^{-1}A_{21} = \begin{bmatrix} 1 & -1 \\ 0 & -1 \end{bmatrix},
\bar{A}_{11} = A_{11} - A_{12}A_{22}^{-1}A_{21} = \begin{bmatrix} 0 & 1 \\ 1 & -1 \end{bmatrix}, \quad \bar{A}_{21} = A_{22}^{-1}A_{21} = \begin{bmatrix} 0 & 1 \end{bmatrix},
\bar{B}_{10} = B_1 - A_{12}A_{22}^{-1}B_2 = \begin{bmatrix} 0 & 1 \\ 1 & -1 \end{bmatrix}, \quad \bar{B}_{11} = E_{12}A_{22}^{-1}B_2 = \begin{bmatrix} 0 & 1 \\ 0 & 1 \end{bmatrix},
\bar{B}_{21} = -A_{22}^{-1}B_2 = \begin{bmatrix} 0 & 1 \end{bmatrix}.$$
(35)

Step 4. Using (30b) and (35) we obtain

$$\hat{A}_{11} = \bar{E}_{11}^{-1} \bar{A}_{11} = \begin{bmatrix} 1 & 0 \\ 1 & -1 \end{bmatrix}, \quad \hat{A}_{21} = -\bar{A}_{21} \bar{E}_{11}^{-1} \bar{A}_{11} = \begin{bmatrix} 1 & 0 \end{bmatrix},$$
$$\hat{B}_{10} = \bar{E}_{11}^{-1} \bar{B}_{10} = \begin{bmatrix} 1 & 1 \\ 1 & 0 \end{bmatrix}, \quad \hat{B}_{20} = -\bar{A}_{21} \bar{E}_{11}^{-1} \bar{B}_{10} = \begin{bmatrix} -1 & 1 \end{bmatrix}, \quad (36)$$
$$\hat{B}_{11} = \bar{E}_{11}^{-1} \bar{B}_{11} = \begin{bmatrix} 0 & 2 \\ 0 & 1 \end{bmatrix}, \quad \hat{B}_{21} = \bar{B}_{21} - \bar{A}_{21} \bar{E}_{11}^{-1} = \begin{bmatrix} 0 & 0 \end{bmatrix}.$$

Therefore, the dynamical part of the fractional discrete-time system with (31) is described by the differential equation

$$\Delta^{\alpha} \begin{bmatrix} x_{1i} \\ x_{2i} \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ 1 & -1 & 0 \\ 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_{1i} \\ x_{2i} \end{bmatrix} + \begin{bmatrix} 1 & 1 \\ 1 & 0 \\ -1 & 1 \end{bmatrix} u_i + \begin{bmatrix} 0 & 2 \\ 0 & 1 \\ 0 & 0 \end{bmatrix} \Delta^{\alpha} u_i , \quad i = 0, 1, \dots$$
(37)

and the static part by the algebraic equation

$$\begin{bmatrix} 0 & 1 \end{bmatrix} x_{1i} + x_{2i} + \begin{bmatrix} 0 & 1 \end{bmatrix} u_i = 0, \quad i = 0, 1, \dots$$
(38)

4. Concluding remarks

A new method of the decomposition of the fractional descriptor linear continuous-time and discrete-time systems into dynamical and static parts has been proposed. Conditions for the decomposition of the fractional descriptor linear systems are established and procedures for decompositions of the matrices into the dynamical and static parts have been given. The procedures have been illustrated by numerical examples. Open problems are extensions of these considerations to the fractional different orders linear systems and to 2-D standard and fractional orders discrete-time linear systems.

References

- L. DAI: Singular Control Systems. Lectures Notes in Control and Information Sciences, Springer-Verlag, Berlin, 1980.
- [2] G.R. DUAN: Analysis and Design of Descriptor Linear Systems. Springer-Verlag, New York, 2010.
- [3] T. KACZOREK: Absolute stability of a class of fractional positive nonlinear systems. *International Journal of Applied Mathematics and Computer Science*, 29(1), (2019), 93–98. DOI: 10.2478/amcs-2019-0007
- [4] T. KACZOREK: Positivity and stability of descriptor linear systems with interval state matrices. *Computational Problems of Electrical Engineering*, 8(1), (2018), 7–17. DOI: 10.23939/jcpee2018.01.007
- [5] T. KACZOREK: Positive linear systems consisting of n subsystems with different fractional orders. *IEEE Transactions on Circuits and Systems*, 58(7), (2011), 1203–1210. DOI: 10.1109/TCSI.2010.2096111
- [6] T. KACZOREK: Selected Problems of Fractional Systems Theory. Springer, Berlin, 2011.
- [7] T. KACZOREK and K. BORAWSKI: *Descriptor Systems of Integer and Fractional Orders*. Springer Nature, Switzerland AG, 2021.
- [8] T. KACZOREK and K. ROGOWSKI: Fractional Linear Systems and Electrical Circuits. Springer, Cham, 2015.
- [9] T. KACZOREK and A. RUSZEWSKI: Analysis of the fractional descriptor discrete-time linear systems by the use of the shuffle algorithm. *Journal of Computational Dynamics*, 8(2), (2021), 153–163. DOI: 10.3934/jcd.2021007
- [10] W. MITKOWSKI: Outline of Control Theory. AGH Publisher, Kraków, 2019, (in Polish).
- [11] P. OSTALCZYK: Discrete Fractional Calculus. World Scientific, River Edgle NJ, 2016.
- [12] I. PODLUBNY: Fractional Differential Equations. Academic Press, San Diego, 1999.
- [13] A. RUSZEWSKI: Practical and asymptotic stabilities for a class of delayed fractional discretetime linear systems. *Bulletin of the Polish Academy of Sciences: Technical Sciences*, **67**(3), (2019), 509–515. DOI: 10.24425/ bpasts.2019.128426
- [14] A. RUSZEWSKI: Stability of discrete-time fractional linear systems with delays. Archives of Control Sciences, 29(3), 2019, 549–567. DOI: 10.24425/acs.2019.130205
- [15] L. SAJEWSKI: Decentralized stabilization of descriptor fractional positive discrete-time linear systems with delays. In: Szewczyk, R., Zieliński, C., Kaliczyńska, M. (Eds.) Automation 2018, Advances in Intelligent Systems and Computing, 743, Springer, Cham, 2018.
- [16] Ł. SAJEWSKI: Stabilization of positive descriptor fractional continuous-time linear system with two different fractional orders by decentralized controller. 21st International Conference on System Theory, Control and Computing (ICSTCC), Sinaia, Romania, (2017), 827–832.