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### Drazin inverse matrix method for fractional descriptor discrete-time linear systems

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Abstract. The Drazin inverse of matrices is applied in order to find the solutions of the state equations of fractional descriptor discrete-time linear systems. The solution of the state equation is derived and the set of consistent initial conditions for a given set of admissible inputs is established. The proposed method is illustrated by a numerical example.

Key words: Drazin inverse, descriptor, fractional, discrete-time, linear system solution.

#### 3. Introduction

Descriptor (singular) linear systems have been considered in many papers and books [1–19]. The eigenvalues and invariants assignment by state and output feedbacks have been investigated in [8, 12, 17], and the minimum energy control of descriptor linear systems in [20, 21]. The computation of Kronecker's canonical form of singular pencil has been analyzed in [19]. The positive linear systems with different fractional orders have been addressed in [22]. Selected problems in theory of fractional linear systems have been described in monograph [16].

Descriptor and standard positive linear systems with the use of Drazin inverse have been addressed in [1–4, 10, 13, 17]. The shuffle algorithm has been applied to checking the positivity of descriptor linear systems in [11]. The stability of positive descriptor systems has been investigated in [23]. Reduction and decomposition of descriptor fractional discrete-time linear systems have been considered in [14]. A new class of descriptor fractional linear discrete-time system has been introduced in [15].

The Drazin inverse for finding the solution to the state equation of fractional continuous-time linear systems has been applied in [10] and the controllability, reachability and minimum energy control of fractional discrete-time linear systems with delays in state have been investigated in [24]. A comparison of three different methods for finding the solution for descriptor fractional discrete-time linear system has been presented in [25].

In this paper the solution to the state equation of fractional descriptor discrete-time linear systems by the use of Drazin inverse of matrices will be derived.

The paper is organized as follows. In section 2 the state equation of the fractional descriptor discrete-time linear systems and some basic definitions of the Drazin inverse and its properties are recalled. The solution to the state equation is presented in section 3 and illustrated with a numerical example in section 4. Concluding remarks are given in section 5.

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The following notation will be used:  $\Re$  – the set of real numbers,  $\Re^{n \times m}$  – the set of  $n \times m$  real matrices and  $\Re^n = \Re^{n \times 1}$ ,  $Z_+$  – the set of nonnegative integers,  $I_n$  – the  $n \times n$  identity matrix, ker A (im A) – the kernel (image) of the matrix.

## 3. Fractional descriptor discrete-time linear systems

Consider the fractional descriptor discrete-time linear system

$$E\Delta^{\alpha} x_{i+1} = A x_i + B u_i, \ i \in \mathbb{Z}_+ = \{0, 1, \dots\},$$
(1)

where  $x_i \in \Re^n$  is the state vector  $u_i \in \Re^m$  is the input vector,  $E, A \in \Re^{n \times n}$ ,  $B \in \Re^{n \times m}$  and

$$\Delta^{\alpha} x_i = \sum_{j=0}^{i} (-1)^j \binom{\alpha}{j} x_{i-j}$$
(2a)

$$\binom{\alpha}{j} = \begin{cases} \frac{1}{\alpha(\alpha-1)\dots(\alpha-j+1)} & \text{for } j=0\\ \frac{j!}{j!} & \text{for } j=1,2,\dots \end{cases}$$
(2b)

is the fractional  $\alpha \in \Re$  order difference of  $x_i$ . Substituting (2) into (1) we obtain

$$Ex_{i+1} = A_{\alpha}x_i + \sum_{j=2}^{i+1} c_j Ex_{i-j+1} + Bu_i$$
(3a)

where

$$A_{\alpha} = A + E\alpha, \ c_j = (-1)^j \binom{\alpha}{j}.$$
 (3b)

It is assumed that det E = 0, but

$$\det[Ez - A_{\alpha}] \neq 0 \text{ for some } z \in \mathbb{C}, \qquad (4)$$

where C is the field of complex numbers.

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Assuming that for some chosen  $c \in \mathbb{C}$ , det $[Ec - A_{\alpha}] \neq 0$  and premultiplying (3a) by  $[Ec - A_{\alpha}]^{-1}$ , we obtain

$$\overline{E}x_{i+1} = \overline{A}_{\alpha}x_i + \sum_{j=2}^{i+1} c_j \overline{E}x_{i-j+1} + \overline{B}u_i , \qquad (5a)$$

where

$$\overline{E} = [Ec - A_{\alpha}]^{-1}E, \ \overline{A}_{\alpha} = [Ec - A_{\alpha}]^{-1}A_{\alpha},$$

$$\overline{B} = [Ec - A_{\alpha}]^{-1}B.$$
(5b)

Note that the equations (3a) and (5a) have the same solution  $x_i, i \in \mathbb{Z}_+$ .

**Definition 1.** [4, 17] The smallest nonnegative integer q is called the index of the matrix  $\overline{E} \in \Re^{n \times n}$  if

$$\operatorname{rank} \overline{E}^{q} = \operatorname{rank} \overline{E}^{q+1}.$$
(6)

**Definition 2.** [4, 17] A matrix  $\overline{E}^{D}$  is called the Drazin inverse of  $\overline{E} \in \Re^{n \times n}$  if it satisfies the conditions

$$\overline{E}\overline{E}^{D} = \overline{E}^{D}\overline{E} , \qquad (7a)$$

$$\overline{E}^{D}\overline{E}\overline{E}^{D} = \overline{E}^{D}, \qquad (7b)$$

$$\overline{E}^{\,D}\overline{E}^{\,q+1} = \overline{E}^{\,q}\,,\tag{7c}$$

where q is the index of  $\overline{E}$  defined by (6). The Drazin inverse  $\overline{E}^D$  of a square matrix  $\overline{E}$  always exists and is unique [4, 17]. If det  $\overline{E} \neq 0$  then  $\overline{E}^D = \overline{E}^{-1}$ . Some methods for computation of the Drazin inverse are given in [17, 19] and in the Appendix.

**Theorem 1.** The matrices  $\overline{E}$  and  $\overline{A}_{\alpha}$  defined by (5b) satisfy the following equalities

1. 
$$\overline{A}_{\alpha}\overline{E} = \overline{E}\overline{A}_{\alpha}$$
 and  $\overline{A}_{\alpha}^{D}\overline{E} = \overline{E}\overline{A}_{\alpha}^{D}$ ,  $\overline{E}^{D}\overline{A}_{\alpha} = \overline{A}_{\alpha}\overline{E}^{D}$ ,  
 $\overline{A}_{\alpha}^{D}\overline{E}^{D} = \overline{E}^{D}\overline{A}_{\alpha}^{D}$ , (8a)

2. 
$$\ker \overline{A}_{\alpha} \cap \ker \overline{E} = \{0\},$$
 (8b)

3. 
$$\overline{E} = T \begin{bmatrix} J & 0 \\ 0 & N \end{bmatrix} T^{-1}, \quad \overline{E}^{D} = T \begin{bmatrix} J^{-1} & 0 \\ 0 & 0 \end{bmatrix} T^{-1}, \quad (8c)$$

det  $T \neq 0$ ,  $J \in \Re^{n_1 \times n_1}$ , is nonsingular,  $N \in \Re^{n_2 \times n_2}$ 

is nilpotent,  $n_1 + n_2 = n$ ,

4. 
$$(I_n - \overline{E}\overline{E}^D)\overline{A}_{\alpha}\overline{A}_{\alpha}^D = I_n - \overline{E}\overline{E}^D$$
 and  
 $(I_n - \overline{E}\overline{E}^D)(\overline{E}\overline{A}_{\alpha}^D)^q = 0.$  (8d)

Proof. Using (5b) we obtain

$$\overline{E}c - \overline{A}_{\alpha} = [Ec - A_{\alpha}]^{-1}[Ec - A_{\alpha}] = I_n$$
(9)

and

$$\overline{A}_{\alpha} = \overline{E}c - I_n \,. \tag{10}$$

Therefore

$$\overline{EA}_{\alpha} = \overline{E}[\overline{E}c - I_n] = [\overline{E}c - I_n]\overline{E} = \overline{A}_{\alpha}\overline{E}.$$
(11)

The proof of the remaining equalities (8) is similar.  $\Box$ Theorem 2. Let

$$P = \overline{E}\overline{E}^{D}, \qquad (12a)$$

and

$$Q = \overline{E}{}^{D}\overline{A}_{\alpha}.$$
 (12b)

Then:

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5. 
$$P^k = P$$
 for  $k = 2, 3, ...$  (13)

$$5. PQ = QP = Q, \tag{14}$$

$$7. \quad PE^D = E^D \,. \tag{15}$$

**Proof.** Using (12a) we obtain

$$P^{2} = \overline{E}\overline{E}^{D}\overline{E}\overline{E}^{D} = \overline{E}\overline{E}^{D} = P$$
(16)

since by (7b)  $\overline{E}^{D}\overline{E}\overline{E}^{D} = \overline{E}^{D}$  and by induction

$$P^{k} = P^{k-1}P = \overline{E}\overline{E}^{D}\overline{E}\overline{E}^{D} = P^{2} = P \text{ for } k = 2,3,\dots$$
 (17)

Using (12) we obtain

$$PQ = \overline{E}\overline{E}{}^{D}\overline{E}{}^{D}\overline{A}_{\alpha} = \overline{E}{}^{D}\overline{E}\overline{E}{}^{D}\overline{A}_{\alpha} = \overline{E}{}^{D}\overline{A}_{\alpha} = Q$$
(18)

and

$$QP = \overline{E}^{D}\overline{A}_{\alpha}\overline{E}\overline{E}^{D} = \overline{E}^{D}\overline{E}\overline{A}_{\alpha}\overline{E}^{D} = \overline{E}^{D}\overline{E}\overline{E}^{D}\overline{A}_{\alpha} = \overline{E}^{D}\overline{A}_{\alpha} = Q$$
(19)

Using (12a), (7a) and (7b) we obtain

$$P\overline{E}^{\,D} = \overline{E}\overline{E}^{\,D}\overline{E}^{\,D} = \overline{E}^{\,D}\overline{E}\overline{E}^{\,D} = \overline{E}^{\,D}\,. \tag{20}$$

#### 3. Solution to the state equation by the use of **Drazin inverse**

In this section the solution to the state equation (1) will be derived by the use of the Drazin inverses of the matrices  $\overline{E}$ and  $\overline{A}_{\alpha}$ .

**Theorem 3.** The solution to the equation (5a) is given by

$$x_{i} = Q^{i} Pv + c_{2} Q^{i-2} Pv + c_{3} Q^{i-3} Pv + \dots + 2c_{i-1} QPv + c_{i} Pv + \sum_{k=0}^{i-1} \overline{E}^{D} Q^{i-k-1} \overline{B} u_{k} + (P - I_{n}) \sum_{k=0}^{q-1} Q^{k} \overline{A}_{\alpha}^{D} \overline{B} u_{i+k}$$
(21)

where Q and P are defined by (12), coefficient  $c_i$  can be computed using (3b) and  $v \in \Re^n$  is arbitrary.

Proof. The system is linear thus the proof can be accomplished independently for the initial conditions and inputs. Taking into account only the first term of (21), we obtain

$$\overline{E}x_{i+1} = \overline{E}[Q^{i+1}Pv + c_2Q^{i-1}Pv + c_3Q^{i-3}Pv + \dots + 2c_iQPv + c_{i+1}Pv] 
= \overline{A}_{\alpha}[Q^iPv + c_2Q^{i-2}Pv + c_3Q^{i-3}Pv + \dots + 2c_{i-1}QPv + c_iPv] (22) 
= \overline{A}_{\alpha}x_i + \sum_{k=2}^{i}c_k\overline{E}x_{i-k+1}$$

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since by Theorems 1 and 2 conditions (8a) and (13 - 15) hold. The proof for the next two terms of (21) is similar to the proof for standard descriptor discrete-time linear systems given in [4, 17].  $\Box$ 

From (21) for i = 0 we have

$$x_0 = Pv + (P - I_n) \sum_{k=0}^{q-1} \mathcal{Q}^k \overline{A}_{\alpha}^D \overline{B} u_k .$$
<sup>(23)</sup>

Equality (23) defines the set of consistent initial conditions  $X_0$ ,  $x_0 \in X_0$  for given a set of admissible inputs  $U_{ad}$ ,  $u_k \in U_{ad}$ , k = 0, 1, ..., q - 1.

If  $u_k = 0$ ,  $k = 0, 1, \dots, q - 1$ , then from (23) we obtain

$$x_0 = Pv \text{ and } x_0 \in \operatorname{Im} P \tag{24}$$

where  $I_{Im P}$  denotes the image of P.

**Remark 1.** The solution to the equation (5a) for  $u_i = 0$ ,  $i \in Z_+$  can be computed recurrently using the formula

$$x_{i} = Qx_{i-1} + \sum_{k=2}^{l} c_{k} Px_{i-k} .$$
(25)

Theorem 2. Let

$$\Phi_0(i) = Q^i + \sum_{k=2}^i c_k \overline{A}_{\alpha} P, \qquad (26)$$

$$\Phi(i) = \sum_{k=0}^{i-1} \overline{E}^D Q^{i-k-1} \overline{B} , \qquad (27)$$

where Q and P are defined by (12). Then

$$P\Phi_0(i) = \Phi_0(i), \qquad (28)$$

$$P\Phi(i) = \Phi(i) . \tag{29}$$

Proof. Using (12) and (26), we obtain

$$P\Phi_0(i) = P\left[Q^i + \sum_{k=2}^i c_k \overline{A}_{\alpha}P\right] = Q^i + \sum_{k=2}^i c_k \overline{A}_{\alpha}P \qquad (30)$$

since (14) and (13) hold. Similarly, using (27), (14) and (15) we obtain

$$P\Phi(i) = \sum_{k=0}^{i-1} P\overline{E}^D Q^{i-k-1}\overline{B} = \sum_{k=0}^{i-1} \overline{E}^D Q^{i-k-1}\overline{B} = \Phi(i).$$
(31)

#### 4. Example

Consider the equation (1) with the matrices

$$E = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix}, \quad A = \begin{bmatrix} 0 & 1 & 0 \\ -2 & -3 & 0 \\ 1 & 2 & -1 \end{bmatrix}, \quad B = \begin{bmatrix} 1 \\ 0 \\ 2 \end{bmatrix} \text{ for } \alpha = 0.5. (32)$$

In this case we have

$$A_{\alpha} = A + E\alpha = \begin{bmatrix} 0.5 & 1 & 0 \\ -2 & -2.5 & 0 \\ 1 & 2 & -1 \end{bmatrix}.$$
 (33)

The pencil of (32) is regular, since

$$\det[Ez - A_{\alpha}] = \begin{vmatrix} z - 0.5 & -1 & 0 \\ 2 & z + 2.5 & 0 \\ -1 & -2 & 1 \end{vmatrix}$$
(34)

We choose c = 0 and the matrices (5b) take the forms

$$\overline{E} = [Ec - A_{\alpha}]^{-1}E = [-A_{\alpha}]^{-1}E$$

$$= \begin{bmatrix} -0.5 & -1 & 0 \\ 2 & 2.5 & 0 \\ -1 & -2 & 1 \end{bmatrix}^{-1} \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 0 \end{bmatrix}$$

$$= \frac{1}{0.75} \begin{bmatrix} 2.5 & 1 & 0 \\ -2 & -0.5 & 0 \\ -1.5 & 0 & 0 \end{bmatrix},$$

$$\overline{A}_{\alpha} = [-A_{\alpha}]^{-1}A_{\alpha} = \begin{bmatrix} -1 & 0 & 0 \\ 0 & -1 & 0 \\ 0 & 0 & -1 \end{bmatrix}.$$
(35)

To compute the Drazin inverse of the matrix  $\overline{E}$  we use the procedure given in the Appendix and we obtain

$$\overline{E} = VW, \quad V = \frac{1}{0.75} \begin{bmatrix} 2.5 & 1 \\ -2 & -0.5 \\ -1.5 & 0 \end{bmatrix}, \quad W = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix},$$
$$\overline{E}^{D} = V[W\overline{E}V]^{-1}W$$
$$= \frac{1}{0.75} \begin{bmatrix} 2.5 & 1 \\ -2 & -0.5 \\ -1.5 & 0 \end{bmatrix} \begin{bmatrix} 7.556 & 3.556 \\ -7.111 & -3.111 \end{bmatrix}^{-1} \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \end{bmatrix} \qquad (36)$$
$$= \begin{bmatrix} -0.5 & -1 & 0 \\ 2 & 2.5 & 0 \\ 3.5 & 4 & 0 \end{bmatrix}.$$

and

$$P = E^{D}E$$

$$= \begin{bmatrix} -0.5 & -1 & 0 \\ 2 & 2.5 & 0 \\ 3.5 & 4 & 0 \end{bmatrix} \frac{1}{0.75} \begin{bmatrix} 2.5 & 1 & 0 \\ -2 & -0.5 & 0 \\ -1.5 & 0 & 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 1 & 2 & 0 \end{bmatrix}, \quad (37a)$$

$$Q = \overline{E}^{D} \overline{A}_{\alpha}$$

$$= \begin{bmatrix} -0.5 & -1 & 0 \\ 2 & 2.5 & 0 \\ 3.5 & 4 & 0 \end{bmatrix} \begin{bmatrix} -1 & 0 & 0 \\ 0 & -1 & 0 \\ 0 & 0 & -1 \end{bmatrix} = \begin{bmatrix} 0.5 & 1 & 0 \\ -2 & -2.5 & 0 \\ -3.5 & -4 & 0 \end{bmatrix}.$$
 (37b)

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Using (21) for  $u_k = 0$ , k = 0,1,... and (37) for  $v = \begin{bmatrix} 1 & 2 & 0 \end{bmatrix}^T$  we can compute

$$x_{0} = Pv = \begin{bmatrix} 1\\ 2\\ 5 \end{bmatrix}, \quad x_{1} = Qx_{0} = \begin{bmatrix} 2.5\\ -7\\ -11.5 \end{bmatrix},$$

$$x_{2} = Qx_{1} + c_{2}Px_{0} = \begin{bmatrix} -5.688\\ 12.625\\ 19.563 \end{bmatrix},$$

$$x_{3} = Qx_{2} + c_{2}Px_{1} + c_{3}Px_{0} = \begin{bmatrix} 9.977\\ -20.547\\ -31.117 \end{bmatrix},$$

$$x_{4} = Qx_{3} + c_{2}Px_{2} + c_{3}Px_{1} + c_{4}Px_{0} = \begin{bmatrix} -15.789\\ 31.984\\ 48.18 \end{bmatrix},$$

$$x_{5} = Qx_{4} + c_{2}Px_{3} + c_{3}Px_{2} + c_{4}Px_{1} + c_{5}Px_{0} = \begin{bmatrix} 24.58\\ -49.324\\ -74.068 \end{bmatrix}.$$
(38)

#### 5. Concluding remarks

The Drazin inverse of matrices has been applied to find the solutions of the state equations of the descriptor fractional discrete-time systems with regular pencils. The equality (23) defining the set of admissible initial conditions for given inputs has been derived. Some properties of the matrices P, Q,  $\Phi_0(i)$  and  $\Phi(i)$  have been established (Theorem 2 and 4). The proposed method has been illustrated by a numerical example.

Comparing the presented method with the method based on the Weierstrass decomposition of the regular pencil [16], we may conclude that the proposed method is computationally attractive since the Drazin inverse of matrices can be computed by the use of well-known numerical methods [17, 19]. The presented method can be extended to the positive descriptor fractional continuous-time linear systems. An open problem is an extension of the considerations for standard and positive continuous-discrete descriptor fractional linear systems.

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# A. Procedure for computation of Drazin inverse matrices

To compute the Drazin inverse  $\overline{E}^{D}$  of the matrix  $\overline{E} \in \Re^{n \times n}$  defined by (7b) the following procedure is recommended.

#### **Procedure A.1.**

Step 1. Find the pair of matrices  $V \in \Re^{n \times r}$ ,  $W \in \Re^{r \times n}$  such that

$$\overline{E} = VW$$
, rank  $V = \operatorname{rank} W = \operatorname{rank} \overline{E} = r$ . (A1)

As the *r* columns (rows) of the matrix V(W) the *r* linearly independent columns (rows) of the matrix  $\overline{E}$  can be chosen.

Step 2. Compute the nonsingular matrix

$$W\overline{E}V \in \Re^{r \times r} . \tag{A2}$$

Step 3. The desired Drazin inverse matrix is given by

$$\overline{E}^{D} = V[W\overline{E}V]^{-1}W.$$
(A3)

**Proof.** It will be shown that the matrix (A3) satisfies the three conditions (7) of Definition 2. Taking into account that det  $WV \neq 0$  and (A1) we obtain

$$[W\overline{E}V]^{-1} = [WVWV]^{-1} = [WV]^{-1}[WV]^{-1}.$$
 (A4)

Using (7a), (A1) and (A4) we obtain

$$\overline{E}\overline{E}^{D} = VWV[W\overline{E}V]^{-1}W = VWV[WV]^{-1}[WV]^{-1}W$$

$$= V[WV]^{-1}W$$
(A5a)

and

$$\overline{E}^{D}\overline{E} = V[W\overline{E}V]^{-1}WVW = V[WV]^{-1}[WV]^{-1}WVW$$
  
=  $V[WV]^{-1}W.$  (A5b)

Therefore, the condition (7a) is satisfied. To check the condition (7b) we compute

$$\overline{E}^{D}\overline{E}\overline{E}^{D} = V[W\overline{E}V]^{-1}WVWV[W\overline{E}V]^{-1}W$$

$$= V[WVWV]^{-1}WVWV[W\overline{E}V]^{-1}W$$

$$= V[W\overline{E}V]^{-1}W = \overline{E}^{D}.$$
(A6)

Therefore, the condition (7b) is also satisfied. Using (7c), (A1), (A3) and (A4) we obtain

$$\overline{E}^{D}\overline{E}^{q+1} = V[W\overline{E}V]^{-1}W(VW)^{q+1}$$

$$= V[WV]^{-1}[WV]^{-1}WVW(VW)^{q}$$

$$= V[WV]^{-1}W(VW)^{q} = VW(VW)^{q-1}$$

$$= (VW)^{q} = \overline{E}^{q}$$
(A7)

where q is the index of  $\overline{E}$ .

Therefore, the condition (7c) is also satisfied.