

# Design of an Ultrawideband CPW-Fed Monopole Antenna with a Band-Notch Function

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**Abstract**—In this paper, a CPW-fed monopole antenna for ultra wideband (UWB) applications using band-notch characteristics with size of  $(24 \times 24 \times 1.6 \text{ mm}^3)$  is presented. The antenna is designed for operation across the entire UWB from 3.1 to 10.6 GHz with band-notch in 5-6 GHz, by inserting an H-shaped slot on the CPW staircase feed line. The proposed antenna has a reflection coefficient below -10 dB through the frequency band. Also, reasonable gain values and good radiation pattern over the same frequency band have been observed. Details of the proposed antenna and experimental results are discussed and parametric study is presented.

**Keywords**—Monopole antenna, notch-band, coplanar waveguide (CPW), UWB.

## I. INTRODUCTION

ACCORDING to high demand of the high data transmitting rate, low power consumption and small size, ultra-wideband (UWB) antennas have received great attention for wireless communication applications [1]. With regards to the FCC regulations, the frequency band from 3.1 to 10.6 GHz can be used for high speed wireless communication system [2]. A coplanar waveguide (CPW)-fed antenna has been reported for ultra-wideband operations [3]. A coplanar waveguide (CPW) monopole antenna becomes popular due to wide operation bandwidth, good radiation pattern, simple structure, light weight and ease of fabrication [4].

In [5], a CPW-fed rectangular slot antenna with a wide bandwidth of 98% over a 6 dB average gain was presented. Another CPW-fed monopole antenna was presented in [6] with tapered ground plane which is suitable for UWB applications. A novel printed monopole antenna fed by a CPW with frequency bandwidth from 3 to 11.3 GHz was obtained in [3]. And in [7] new compact CPW-fed rectangular antenna with a patched inverted-L stub was introduced.

As a matter of fact UWB transmitters should not cause any electromagnetic interference to nearby communication system such as the wireless local area network (WLAN) applications, for this reason we should pay attention to this concept for avoiding interfering with nearby communication systems such as IEEE 802.11a in the U.S. (5.15-5.35 GHz, 5.725-5.825 GHz) and HIPERLAN/2 in Europe (5.15-5.35 GHz, 5.47-5.725 GHz). Therefore UWB antennas with band-notched characteristics in the WLAN frequency band are required.

Some papers have studied effect of coupling between the feed line and the aperture by using different shape of slots (i.e., circular, elliptical, hat-shaped arc-shaped slot) on the

feed line [8]–[13]. In [14] a corner-notched rectangular patch which is loaded by a square back-patch was presented. When it comes to CPW-Fed UWB antennas, most of the notched band characteristics are implemented by cutting slots in the radiation element.

In this paper, a novel UWB CPW-fed monopole antenna with a band-notch function in 5-6.1 GHz and compact size of  $24(L) \times 24(W) \times 1.6 \text{ mm}^3$  is presented for UWB application which covers frequency bandwidth of 3.1-11.6 GHz. The proposed antenna has a reasonable average gain and good radiation pattern. Details of antenna design, simulated and measured results are presented and discussed.

Section II presents the geometry of the proposed antenna. Then in section III the effect on the impedance bandwidth, full band and band-notched function design are analyzed. After that the experimental results including measured return losses and radiation patterns are presented in section IV. Section V gives the conclusions.

## II. ANTENNA CONFIGURATION

The proposed UWB CPW-fed monopole antenna with a band-notch function is shown in Fig. 1. The antenna is printed on FR4 substrate with thickness of 1mm and relative permittivity of 4.4 and loss tangent 0.02. The proposed antenna consists of a square substrate which a square slot is etched at the center of the ground plane.

A CPW-fed line is a staircase strip and two gaps of width  $g$  which are located between the ground plane and a CPW-fed line. The gaps between the strips and ground plane are adjusted to reach reasonable impedance matching.

In proposed CPW-fed monopole antenna, two grounded semi-triangles metallic strips placed at down corners of the square slot and two circular metallic strips placed at up corners of the square slot which is modified ground plane to achieve a wide bandwidth.

Each of the semi-triangle metallic strips has a length of  $T_1$  and  $T_2$  in outside and a length of  $T_3$  and  $T_4$  in inside and each of circular metallic strips has a radius of  $A$  with center position of  $P_1$  and  $P_2$ , H-shape slot has a width of  $W_6$ ,  $W_7$  and length of  $L_6$ ,  $L_7$ ,  $L_8$ . The antenna is excited by the  $50\Omega$  CPW feed line, which was in turn connected to a coaxial cable through a standard  $50\Omega$  SMA connector which is for measurement purpose. The lengths of sides of semi-triangles and center position of circular metallic strips are optimized by using a HFSS EM simulator from ANSOFT.

As well by varying the size of the monopole and two symmetry circular strips on the ground plane, the wideband

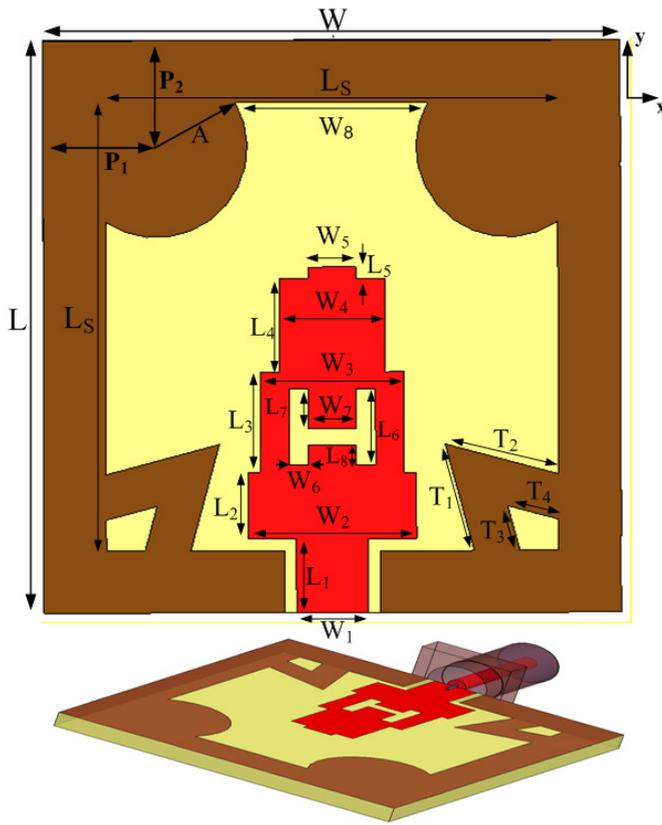


Fig. 1. Geometry of CPW-fed monopole antenna.

operation of the proposed antenna can be excited with a good impedance matching. By cutting the H-shaped slot on the radiating patch an additional surface current path along the slot is created which leads to the band notch characteristic.

The CPW feed line has a staircase shape with widths of  $W_1, W_2, W_3, W_4, W_5$  which are respectively 3, 7, 6, 4.4, 2, 0.8, 2 (units are in mm), and lengths of  $L_1, L_2, L_3, L_4, L_5$  which are respectively 2.6, 2.8, 4.2, 3.9, 0.5 (units are in mm). The inner of values of square slot with dimensions of  $L_s \times L_s$  are  $18.8 \times 18.8 \text{ mm}^2$ .

In proposed CPW-fed monopole antenna, the lengths of each semi-triangles,  $T_1, T_2, T_3$  and  $T_4$  are 4.6 mm, 4.9 mm, 1.93 mm and 2.18 mm respectively,  $A=3.7$  mm and gap distance is  $g=0.5$  mm. The H-shaped slot has width of  $W_6=0.8$  mm,  $W_7=2$  mm, and lengths of  $L_6=3.2$  mm,  $L_7=1.67$  mm and  $L_8=0.85$  mm. The prototype of the antenna is shown in Fig. 1. with

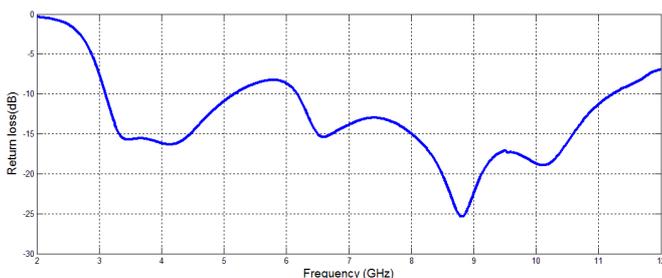


Fig. 2. Simulated reflection coefficient of the proposed antenna.

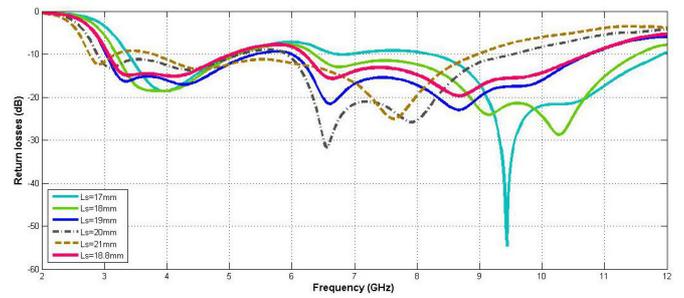


Fig. 3. Simulated reflection coefficients of the proposed antenna with different  $L_s$  with a fixed values of  $P_2=4.5$  and  $T_2=4.87$  (Units are in mm).

$L_s=18.8$  mm,  $P_1=4.4$  mm,  $P_2=4.8$  mm,  $A=3.7$  mm,  $T_2=4.76$  mm has been fabricated and tested.

### III. ANTENNA STRUCTURE STUDIES

In this section, the monopole antenna with various design parameters is constructed. It is clearly seen that the monopole antenna exhibits a broad impedance bandwidth of 8 GHz (3-11 GHz). However, by inserting an H-slot on the staircase feed line one notched band at 5.5 GHz is obtained.

The simulated return loss of the ( $24 \times 24 \text{ mm}^2$ ) CPW-fed monopole antenna with a band notch function is shown in Fig. 2. The result shows that the area of proposed antenna is reduced by 19.6% in comparison with [7], which is  $32.4 \times 24 \text{ mm}^2$ . According to simulation result, the proposed antenna is covered impedance bandwidth of 3.1-5 GHz and 6.1-10.6 GHz for a return loss less than 10 dB.

Simulated reflection coefficients of the proposed antenna with different length of  $L_s$  are shown in Fig. 3. The results show that by increasing the length of  $L_s$  from 17 mm to 21 mm a lower frequency and upper frequency of bandwidth shift downward.

It shows that by increasing the length of  $L_s$ , the size and the area of ground plane will be decreased, consequently lower and upper edge from impedance bandwidth will decrease and by increasing the size of ground plane, impedance bandwidth will increase. The bandwidth is optimized at  $L_s=18.8$  mm.

Similarly, the notch frequency can be obtained by changing the value of  $P_2$  (center of circular grounded strip) on the corner of ground plane from  $P_2=3.8$  mm to 5 mm with a fixed value of  $P_1=4.4$  mm. As shown in Fig. 4, increase the return loss is observed at notched frequency by increasing  $P_2$ . By changing

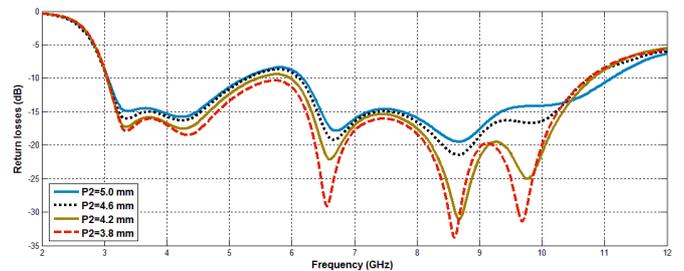


Fig. 4. Simulated reflection coefficients of the proposed antenna with different value of  $P_2$  with a fixed values of  $P_1=4.4$ ,  $L=18.8$  and  $T_2=4.8$  (Units are in mm).

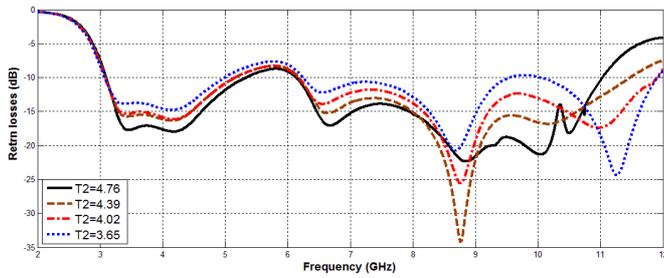


Fig. 5. Simulated reflection coefficients of the proposed antenna with different size of  $T_2$  with a fixed values of  $P_1=4.4$ ,  $P_2=4.5$  and  $L_s=18.8$  (Units are in mm).

the position of circular grounded strip the area of ground plane changed from  $298 \text{ mm}^2$  to  $310 \text{ mm}^2$ .

Additionally simulated reflection coefficients in different side of triangle metallic strips from  $T_2=3.65 \text{ mm}$  to  $4.76 \text{ mm}$ , is shown in fig. 5. It can be seen that notched frequency and lower frequency of bandwidth is almost fixed.

The edge of upper frequency of bandwidth decreased from  $12 \text{ GHz}$  to  $11 \text{ GHz}$  as  $T_2$  increased from  $3.65 \text{ mm}$  to  $4.76 \text{ mm}$ . In this case the area of the ground plane is changed from  $293 \text{ mm}^2$  to  $303 \text{ mm}^2$ .

The gap between radiation patch and ground plane has an important effect on the impedance matching of the proposed antenna, as shown in Fig. 6. When  $g$  is increasing the impedance matching can be greatly improved.

The results show that changing the size of ground plane in all four figures (Fig. 3-Fig. 6) has a direct effect on its area as the area was nearly around  $300 \text{ mm}^2$ , bandwidth of the antenna were almost one and the same.

The geometry of antenna with four grounded semi-triangles metallic strips is shown in Fig. 7. Two circles show in this figure specify the location of circular grounded strip.

Fig. 8 shows the effect of the antenna using circular grounded strip as shown in Fig. 1. (antenna I) in comparison with the antenna using four semi-triangle grounded strip (without circular grounded strip) as shown in Fig. 7. at  $L_s=18.8 \text{ mm}$  (antenna II). It can be seen that in the case of antenna I, circular grounded strip on ground plane cause to achieve a broadband width.

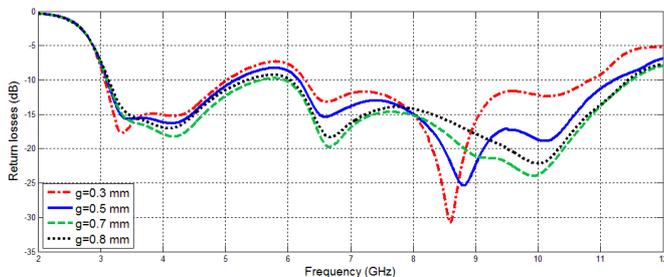


Fig. 6. Simulated reflection coefficients of the proposed antenna with different value of gap distance ( $g$ ) with a fixed values of  $P_1=4.4$ ,  $P_2=4.5$ ,  $L_s=18.8$  and  $T_2=4.8$  (Units are in mm).

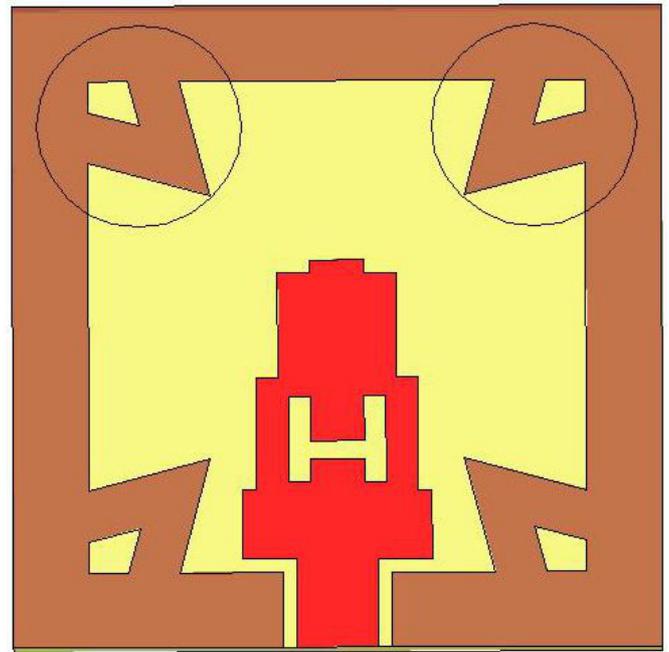


Fig. 7. Geometry of the antenna with four semi-triangles metallic strip in comparison of proposed antenna with two circular grounded strip.

#### IV. EXPERIMENTAL RESULTS

The prototype of the proposed antenna is shown in Fig. 9. The measured reflection coefficients of the proposed antenna ( $L_s=18.8 \text{ mm}$ ,  $P_1=4.4 \text{ mm}$ ,  $P_2=4.8 \text{ mm}$ ,  $A=3.7 \text{ mm}$ ,  $T_2=4.76 \text{ mm}$ ) using an hp 8510 network analyzer in comparison with simulated reflection coefficients is shown in Fig. 10. The antenna bandwidth covers the range of  $3 \text{ GHz}$  to  $11 \text{ GHz}$  with the band-notched function in wireless LAN frequency band.

The measured peak antenna gain is shown in Fig. 11. Clearly, the peak antenna gain is varied from about  $2 \text{ dBi}$  to  $3.5 \text{ dBi}$  across the UWB operating frequencies ( $3$  to  $11 \text{ GHz}$ ). The maximum gain for the antenna is  $3.7 \text{ dBi}$  at  $11 \text{ GHz}$ .

The far-field radiation patterns in the  $xz$ - and  $yz$ -planes at four different frequencies of  $4$ ,  $7$ ,  $8$  and  $10 \text{ GHz}$  for this obtained CPW-fed monopole antenna were also measured and shown in Fig. 12. The patterns in the  $x$ - $y$  plane are all nearly omni-directional, and those in the  $y$ - $z$  plane, as expected, are all very monopole-like and become more directional with the increase of frequency.

Also, the values of cross-polarization in the direction of

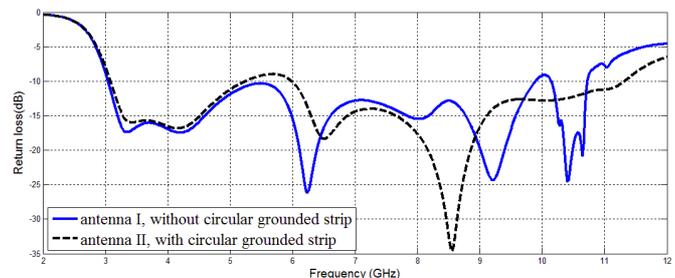


Fig. 8. Simulated reflection coefficient of the antenna I and antenna II, with  $L_s=18.8$ .

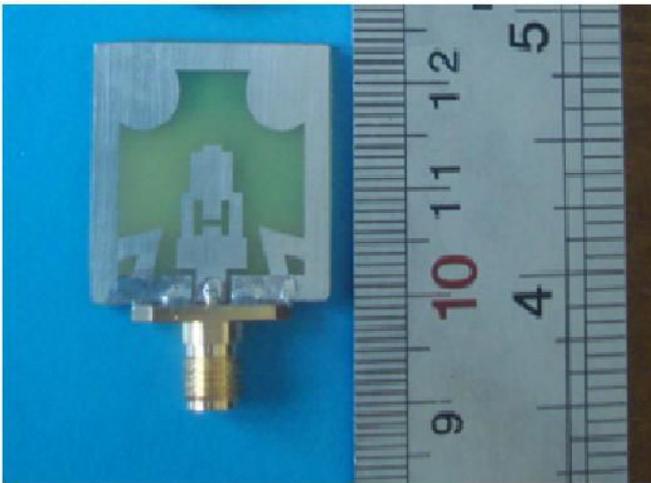


Fig. 9. Prototype of the proposed antenna.

the maximum radiation pattern towards the co-polarization are nearly -20 dB. Furthermore the antenna has stable radiation patterns over the UWB and Good agreements are obtained between the measurement and simulation results.

The simulated current distribution of the proposed antenna at 5.45 GHz is shown in Fig. 13. It reveals that the currents mainly concentrate over the area of H-shaped slot.

By etching an H-shaped slot in the radiating patch of antenna I, a frequency band notch is created. In general terms, the design concept of the notch function is to adjust the total length of the H-shaped slot to be approximately lower than half-wavelength at the desired notched frequency. The current distribution flows around the H-shaped slot. In this case, interference for the excited surface currents in the antenna causes the antenna to show large reflection and band-reject characteristic at that frequency.

## V. CONCLUSION

In this paper, a compact notch-band CPW-fed monopole antenna has been presented. This antenna is fed by staircase CPW line. Two semi-triangles metallic strips are located at down corner and two circular metallic strips are located at up corner of square slot. By using two circular grounded strip the impedance matching in high frequency band can be improved. Proposed antenna is covered frequency bandwidth of 3.1-5

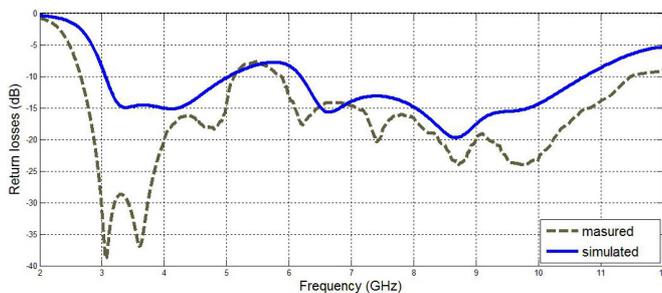


Fig. 10. Simulated and measured reflection coefficients of the proposed antenna with values  $P_1=4.4\text{mm}$ ,  $P_2=4.8\text{mm}$  and  $L_s=18.8\text{mm}$ .

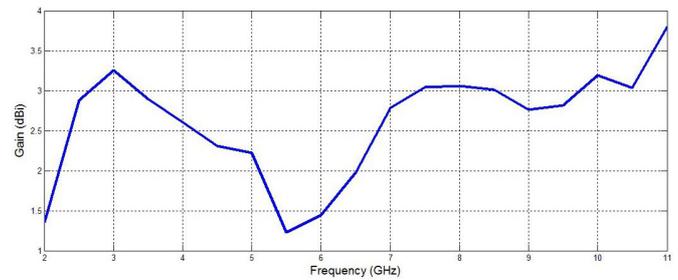


Fig. 11. Measured gain of CPW-fed monopole antenna.

GHz and 6.1-10.6 GHz, also good radiation characteristics are obtained from 3.1 GHz to 10.6 GHz. The proposed antenna is designed on FR4 substrate and the dimension of the antenna is reduced significantly ( $24\times 24\text{ mm}^2$ ). The size reduction is about 20% compared with [7]. The measured maximum peak antenna gain at 10.5 GHz is 11.3 dBi. Therefore, the proposed antenna is suitable for UWB applications simultaneously without interference to nearby communication system such as the wireless local area network (WLAN) applications.

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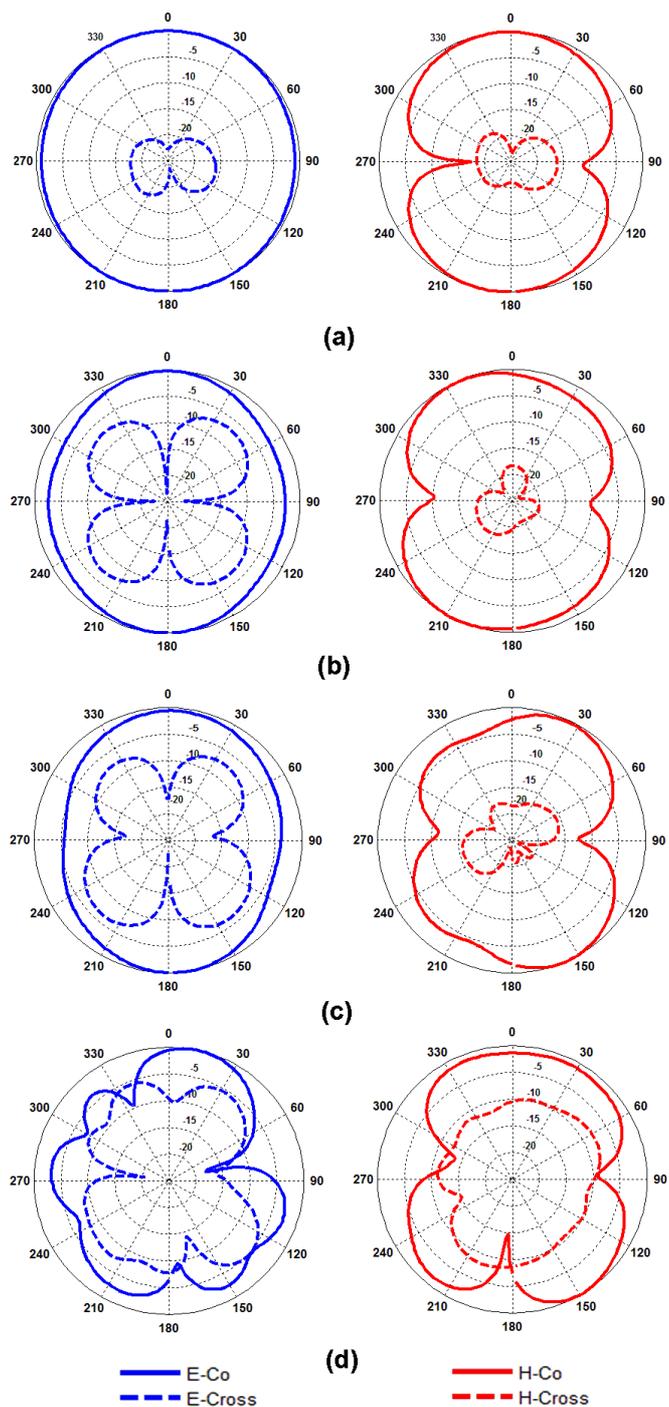


Fig. 12. Measured far field radiation patterns of the proposed antenna at: (a) 4 GHz, (b) 7 GHz (c), 8 GHz, (d) 10 GHz.

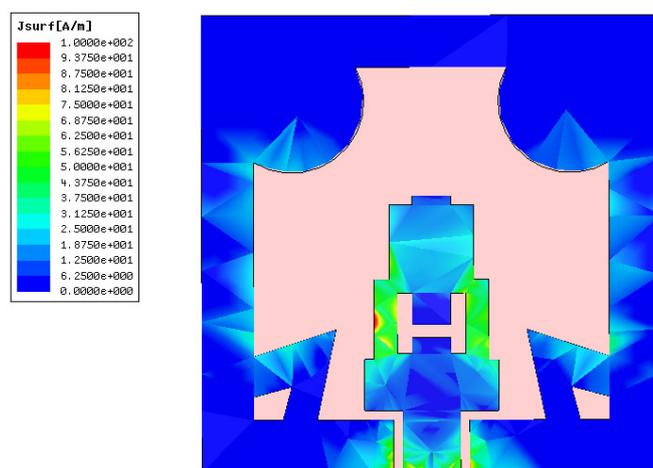


Fig. 13. Simulated current distribution of the proposed antenna at 5.45 GHz.